

L.Malathi, A.Bharathi, A.N.Jayanthi



Abstract—FFT architecture is the common and very efficient design in modern signal much of applications. Though so architectures executed in now-a-days applications, This paper give different approach of FFT design. In order to reduce the computation time, FFT structure is modified in the arrangement. This analyzed approach somewhat satisfies the low power, high performance and to useful in image, signal and wireless applications.

Keywords—FFT, Scaling Factor, Multiplier, CSA, CLA, BM, AM

I. INTRODUCTION

FFT is basically a composition of mathematical analysis. In an invariant approach transformation will estimate the scaling and rotation components. In FFT computation multipliers are play the major role. In the proposed approach there is research is done with the multipliers with the existing results which has done already. Throughout this proposal the high consideration in multipliers and adders used in FFT architecture.

II. METHODOLOGY

A. Survey on FFT Methodology

X(k) is the transformed version of x(n) is mathematically given as,

$$X(k) = \sum_{n=0}^{N-1} x(n) e^{-i2\pi nk/N}, \quad k = 0, 1, \dots, N-1$$
 (1)

$$X(k) = \sum_{n=0}^{N-1} x(n)W_N^{nk}, \quad k = 0, 1, \dots, N-1$$
(2)

Scaling factor for N as common,

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* Correspondence Author

L.Malathi*, Assistant Professor, (Research Scholar of ANNA University, Chennai), Department of Electronics and Communication Engineering, Sri Ramakrishna Institute of Technology, Coimbatore, India, Email:lmalathigraj@gmail.com, ORCID:0000-0002-1558-9431

Dr. A.Bharathi, Professor & Head, Department of Information Technology, Bannari Amman Institute of Technology, Erode, India, Email: bharathia@bitsathy.ac.in

Dr.A.N.Jayanthi, Associate Professor, Department of Electronics and Communication Engineering, Sri Ramakrishna Institute of Technology, Coimbatore, India, Email: jayanthi.ece@srit.org

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$$W_N = e^{-i2\pi/N}$$

$$= \cos\left(\frac{2\pi}{N}\right) - i\sin\left(\frac{2\pi}{N}\right)$$
(3)

$$W_N^{mN} = \left(e^{-i2\pi/N}\right)^{mN}, \quad m = -\infty, \dots, -1, 0, 1, \dots, \infty$$

= $e^{-i2\pi m}$
= 1. (4)

IFFT of X(k) is mathematically given as,

$$x(n) = \frac{1}{N} \sum_{k=0}^{N-1} X(k) e^{i2\pi nk/N}, \quad n = 0, 1, \dots, N-1$$
$$= \frac{1}{N} \sum_{k=0}^{N-1} X(k) W_N^{-nk}, \quad n = 0, 1, \dots, N-1.$$
 (5)

$$X(k) = \sum_{n=0}^{N-1} x(n) W_N^{nk}, \quad k = 0, 1, \dots, N-1$$

$$W_N = e^{-i2\pi/N}$$
(6)

$$X(k) = \sum_{n_{\text{oven}}=0}^{N-2} x(n) W_N^{nk} + \sum_{n_{\text{odd}}=1}^{N-1} x(n) W_N^{nk}.$$
(7)

When implement further we get,

$$X(k) = \sum_{m=0}^{N/2-1} x(2m)W_N^{2mk} + \sum_{m=0}^{N/2-1} x(2m+1)W_N^{(2m+1)k}$$

$$= \sum_{m=0}^{N/2-1} x(2m) (W_N^2)^{mk} + \sum_{m=0}^{N/2-1} x(2m+1) (W_N^2)^{mk} W_N^k.$$
(8)

Scaling factor is,

$$W_N^2 = \left(e^{-i2\pi/N}\right)^2$$

= $e^{-i2\pi 2/N}$
= $e^{-i2\pi/(N/2)}$
= $W_{N/2}$. (9)

Scaling factors actually drive the entire system of FFT throghout from input to output. Hence, delay also an impact on the implentaion result. By analysing the timing behaviour of the scaling factor implementation result will be under on an expected circle.



(10)

(11)

(13)

x(5)

$$\begin{split} X(k) &=& \sum_{m=0}^{N/2-1} x(2m) W_{N/2}^{mk} + W_N^k \sum_{m=0}^{N/2-1} x(2m+1) W_{N/2}^{mk} \\ &=& \sum_{m=0}^{N/2-1} x_{\text{even}}(m) W_{N/2}^{mk} + W_N^k \sum_{m=0}^{N/2-1} x_{\text{odd}}(m) W_{N/2}^{mk}, \\ &k = 0, 1, \dots, N-1 \end{split}$$

$$X(k) = \operatorname{DFT}_{N/2} \left\{ x_{\operatorname{even}}(m), k \right\} \; + \; W_N^k \cdot \operatorname{DFT}_{N/2} \left\{ x_{\operatorname{odd}}(m), k \right\}$$

$$W_N^{x+N/2} = W_N^x W_N^{N/2}$$

$$= W_N^x \left(e^{-i2\pi/N} \right)^{N/2}$$

$$= W_N^x e^{-i2\pi N/2N}$$

$$= W_N^x e^{-i\pi}$$

$$= -W_N^x.$$
(12)

$$X(0) = x(0) + x(1)$$

 $X(1) = x(0) - x(1)$.

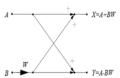


Fig. 1 Radix-2 Butterfly Structure

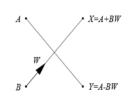
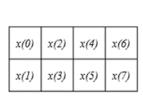


Fig.2 Simplified Radix-2 Butterfly Structure



x(0) x(1) x(2) x(3)

Fig.3 2-D approach of Radix-2 Structure

Fig.4 3-D approach of Radix-2 Structure

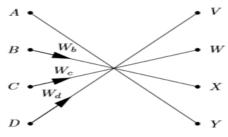


Fig.5 Radix-4 Butterfly Structure

Fig.1 and 2 shows the Radix-2 butterfly structure. Both structures are same but arrangement is different.

Three dimensional and two dimensional approach is implemented in FFT structure as shown in Fig.3 and Fig.4.

The outputs are,

$$V = A + BW_b + CW_c + DW_d$$

$$W = A - iBW_b - CW_c + iDW_d$$

$$X = A - BW_b + CW_c - DW_d$$

$$Y = A + iBW_b - CW_c - iDW_d.$$
(14)

In Eqn, (14) V,W,X,Y are the outputs. V is the addition of A,B,C,D. With B,C,D scaling factor is multiplied. In this step error tolerant multiplier approach is used. Her in order to eliminate the error truncated structure is implemented.

$$B' = BW_b$$

$$C' = CW_c$$

$$D' = DW_d,$$
(15)

Truncated outputs of B,C,D is given in Eqn. 15. Likewise W,X,Y is implemented, In between adders and subtractors are used. In the adder adder design selectors used to choose adder and subtractor.

$$V = A + B' + C' + D'$$

$$W = A - iB' - C' + iD'$$

$$X = A - B' + C' - D'$$

$$Y = A + iB' - C' - iD'.$$
(16)

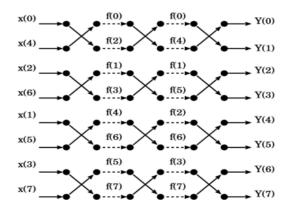


Fig. 6 FFT (8-Bit)

Table. 1 Comparison of FFT Radix

FFT Radix	Number of Complex Multiplications Required		
2	$0.5000 \ MN - (N-1)$		
4	$0.3750 \ MN - (N-1)$		
8	0.3333 MN - (N-1)		
16	$0.3281 \ MN - (N-1)$		

Table. 2 Comparison of FFT with 'N'

Transform length (N)	DFT operations	FFT operations	DFT ops \div FFT ops
16	256	64	4
128	16,400	896	18
1024	1.05×10^{6}	10,240	102
32,768	1.07×10^{9}	4.92×10^{5}	2185
1,048,576	1.10×10^{12}	2.10×10^{7}	52,429





From the table.2 according to N when DFT needs 256 operations, 64 operations as reduced in FFT. If N increases comparatively operations of FFT is reduced than DFT. In according with that the speed is a factor with increasing of N value. The speed in connected parameter with timing. In the proposed structure timing analysis done by in between each node in the entire structure.

Table. 3 Comparison of FFT with

N	FFT Radix	Real Multiplies Required	Real Additions Required
256	2	4096	6144
256	4	3072	5632
256	16	2560	5696
512	2	9216	13,824
512	8	6144	12,672
4096	2	98,304	147,456
4096	4	73,728	135,168
4096	8	65,536	135,168
4096	16	61,440	136,704

Power and Delay analysis of FFT in MOS,

$$P_{Total} = P_{Short-Circuit} + P_{Leakage} + P_{Switching} + P_{Constant-Current}. \\$$

(17)

$$P_{switching} = C_{load} V_{dd}^2 f$$
 W (18)

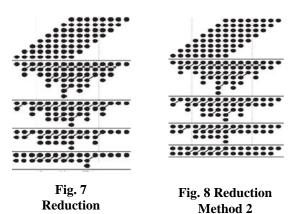
$$P_{switching_{node}} = a C_{load} V_{dd}^2 f W$$
(19)

$$C_{total} = C_{gate} + C_{diffusion} + C_{interconnect}$$
 (20)

$$Delay \approx C_{total}/I$$
 (21)

B. Review on Multipliers

Multipliers have large area and large propogation delays and thereby consume power. Therefore, Reduction in multiplications will give high impact on throughput of FFT output. For that different multipliers were reviewed. In that some multiplier approach is used to calculate the FFT.



Error tolerant multiplier design shown in Fig.9 which depicts that, it includes two different multiplier connected as

parallel manner for MSB and LSB of input. The product also out from the respective multipliers.

An 8-Bit Multiplier Urdhva Tiryakbhyam Sutra is shown in Fig.10 to implement FFT. In that four 2-bit multipliers are connected in parallel manner. For this input is connected as bit serial order. The parallel outputs P0 to P7 out from the architecture.

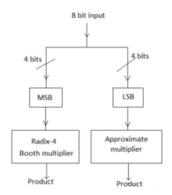


Fig.9 Error Tolerant Multiplier Design

For and addition carry look ahead adder is used. In Fig.11 Look Ahead approach is depicted. In the X(k) the output of FFT, the carry bypass adder is implemented as shown in Fig. 12 in order to bypass the carry out. In Fig.11 First level three AND gates are used to bypass the input to the next level. In second level half adder is used to originate the output. In carry bypass adder eight full adder is used.

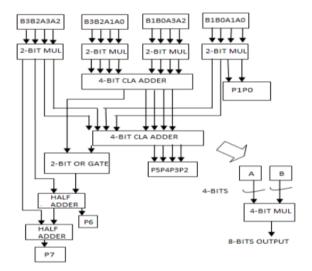


Fig.10 An 8-Bit Multiplier Urdhva Tiryakbhyam Sutra

C. Review on Adders

Adders are the basic unit in multipliers for high accuracy. If the performance of the multiplier is analyzed based on adders, then the other parameters can obtained as per designers requirement.

Here some of the topologies were analyzed according to VLSI approach (i.e.). Proposed adder is suited in different approach.

$$G_i = A_i.B_i$$
 for i=0,1,2,3 (22)

$$C_i(cout) = G_{i+} P_i C_{i-1}$$
(23)

$$S_i = A_i \oplus B_i C_{i-1} \tag{24}$$

Eqn.22 gives the product of input A and B. i=0,1,2,3. Eqn. 23 and 24 gives sum and carry respectively.

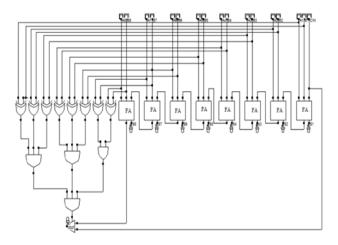


Fig. 11 Look Ahead Approach of Adder

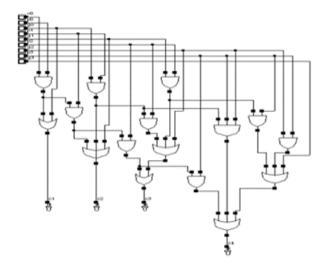


Fig.12 Carry Bypass Adder

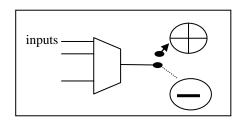


Fig. 13 Selection of Addition or

III. PROPOSED STRUCTURE AND DESIGN ANALYSIS

Optimized proposed here is, [Elisardo Antelo] Low Latency, Reduce Area and Delay, 34x34-digit Multiplication, Area:(20-35)%, [Xiaoxiao Zhang] Reduced Area and Power Consumption, For 32x32 bit,Area:11.1%, [Kostas Tsoumanis] Reduce the Critical Delay, Hardware Complexity and Power Consumption. Power: 2.30%-5.66%,Area: 2.16% - 10.85%, [Shin-Kai Chen] Reduce the Critical Path, Low Latency, [Artur K] Reduced Area, High Speed and High Accuracy, For 16X16-Bit Multiplier,Area:11.4%-25.4%, Reduced Latency and Delay,Latency:447Ps@2.5mV [Reza Azarderakhsh] Reduce the Latency, Reduce Area and Time Delay at Different Digit Sizes, High Speed Computation,By Using AISC and FPGA at Different Digit Size,Reduced Area & High Accuracy,DT Multiplier Error Reduced (87%, 16x16),Transistor Count is reduced by 47%.

A. Simulation Summary

Simulation results from the proposed structure is having acceptable rate of area, delay and power is reduced.

IV. RESULTS

A. Simulation Results

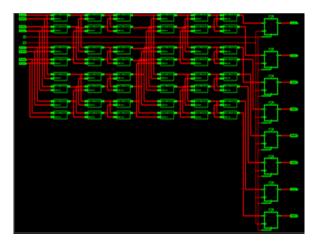


Fig.14 FFT Schematic

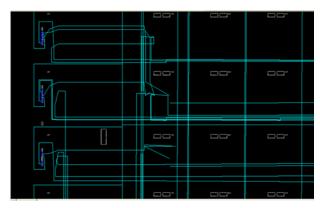


Fig.15 FPGA View -



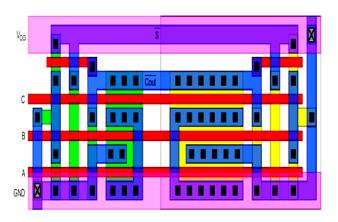


Fig.16 Layout - Adder

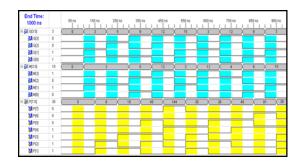


Fig.17 Simulation Result - Adder

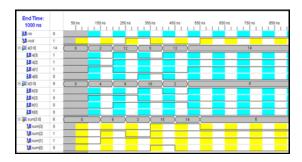


Fig.18 Simulation Result - Multiplier

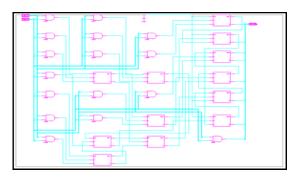


Fig.19 Schematic Multiplier

V. DISCUSSION

Based on the proposed architecture is some of the implementation discussions have to do by applying in finite response structures and multiplier based applications with an increasing bit rate.

VI. CONCLUSION

New dynamic data scaling architecture for pipeline FFT/IFFT have been proposed for 1-D applications. Based on fixed point pipeline low memory and arithmetic requirements have been constructed. This FFT/IFFT structure can be implemented in OFDM applications as Modulation/Demodulation in communications for high signal quality. High throughput outputs have been obtained and scaling of data stalling has been avoided when data is continuously streaming by using pipeline architecture. Instead of DFT, in this project FFT is proposed for speedup the computations. In FFT/IFFT computation, if we increase the size of word length, quality of the signal and the throughput will automatically increased.

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AUTHORS PROFILE



Ms. L. Malathi, She received her B.E. degree in Electronics and Communication Engineering in the year 2005 and M.E. degree in Applied Electronics from Anna University, Chennai in the year 2008. Currently she is working as Assistant Professor (Sr. Gr.) in Department of

Electronics and Communication Engineering at Sri Ramakrishna Institute of Technology, Coimbatore, India. She is having 12 years of teaching experience. She has presented papers in various National and International Conferences & Journals. Her field of interest is VLSI and Digital Signal Processing. She is the Life Time Member ISTE, IAENG and ACM. She is currently pursuing Ph.D. in the area of VLSI, Anna University, Chennai.



Dr. A. Bharathi, She received her Ph.D degree in 2014 in Faculty of Information and Communication Engineering, Anna University, Chennai in the year 2007. She did her Bachelor's degree at Kongu Engineering College, Perundurai in Computer Science and Engineering under

Bharathiar University, Coimbatore in the year 1998. She then completed her Post Graduate Degree at Bannari Amman Institute of Technology in Computer Science and Engineering under Anna University, Chennai, in the year 2007. She finished her Doctoral degree in Information and Communication Engineering specializing in Data Mining. She is the Life Time Member of ISTE. She has over 22 years of teaching experience. She got Awards and Patent. She has delivered Guest Lectures in various institutions.



Dr. A.N. Jayanthi, She received her Ph.D degree in 2014 in Faculty of Information and Communication Engineering, Anna University, Chennai. She received her M.E degree in VLSI Design from Government College of Technology, Coimbatore, Anna university, Chennai in

the year 2008 and her B.E degree in Electronics and Communication Engineering from Sri Ramakrishna Engineering College, Coimbatore under Bharathiar University, Coimbatore in the year 1998. She is having 22 years of teaching experience .Currently she is working as Associate Professor in the Department of Electronics and Communication Engineering at Sri Ramakrishna Institute of Technology, Coimbatore. Her area of specialization in Ph.D. is VLSI Design. She has published many research papers in referred National and International Journals & Conferences. She is a Life Time Member of ISTE.



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